

**ShanghaiTech University**  
**School of Information Science and Technology**

**EE112 Lab Experiments**

**Diode, BJT, MOSFET**

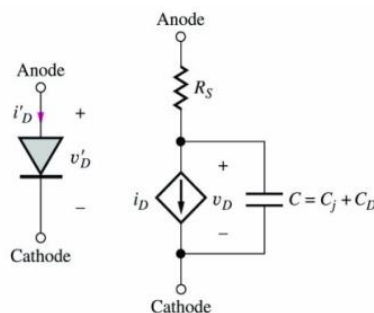
**SPICE Model of pn Junction:**

In class, we described the current-voltage and capacitance-voltage relationship as

$$i_D = I_s \left[ \exp\left(\frac{qv_D}{nkT}\right) - 1 \right] = I_s \left[ \exp\left(\frac{v_D}{nV_T}\right) - 1 \right]$$

$$C_j = \frac{C_{j0}A}{\sqrt{1 + \frac{v_R}{\phi_j}}}$$

The parameter  $n$  is called “ideality factor”. For ideal diode at normal current,  $n$  is equal to 1. In real devices, it could deviate from 1. The ideality factor can be easily extracted from the semi-log plot of  $i_D$  vs.  $V_D$ . The SPICE model of a pn junction is shown below:

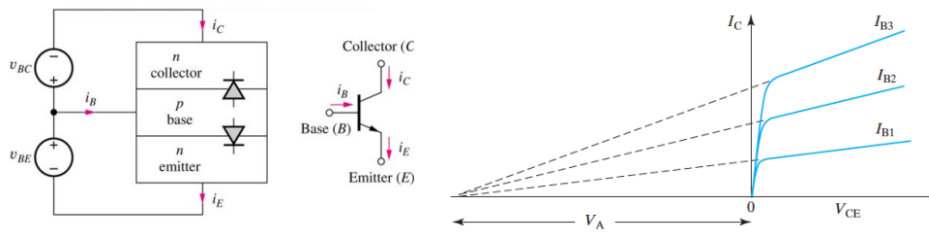


$$i_D = IS \left[ \exp\left(\frac{v_D}{NV_T}\right) - 1 \right]$$

$$C_D = TT \frac{i_D}{NV_T} \text{ for } v_D \geq 0 \quad C_j = \frac{CJO}{\left(1 - \frac{v_D}{VJ}\right)^M} \text{ RAREA for } v_D \leq 0$$

Note the expression for  $C_j$  is slightly different. If  $M = 0.5$ , the expression is exactly the same as that derived in class. This corresponds to pn junctions with constant doping concentrations, i.e.,  $N_D$  and  $N_A$  are constant. This is called “abrupt junction”. If they are not constant, for example, in so-called “graded” pn junctions,  $M$  could deviate from 0.5. The values of  $M$  and built-in potential,  $V_j$  (or  $\phi_j$ ), can be extracted from experimental data by curve fitting. If  $M = 0.5$ ,  $V_j$  can be found by plotting  $1/C_j^2$  vs.  $V_D$ . The SPICE model also includes diffusion capacitance,  $C_D$ . It is negligible for reverse bias. We will ignore it here. The SPICE model also includes a series resistance,  $R_S$ . Its value can be extracted from the slope of the I-V curve at high current: the total resistance at high current is dominated by  $R_S$ .

**Bipolar Junction Transistor (BJT):**



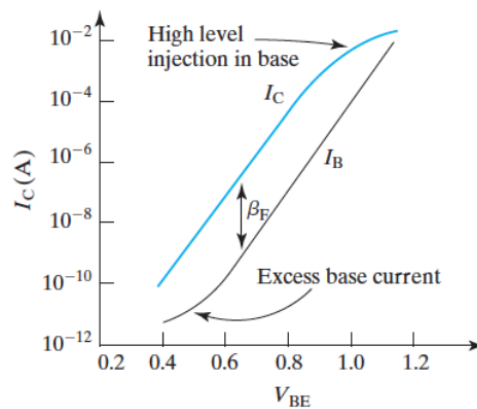
The schematic structure, symbol, and the current-voltage (I-V) characteristics of a BJT are shown above. The collector current,  $I_C$ , is controlled by base current,  $I_B$ . The flat part of the I-V curves is called “forward active” region. The flatness of the  $I_C$  curves is described by “Early voltage”,  $V_A$ :

$$r_0 = \left( \frac{\partial I_C}{\partial V_{CE}} \right)^{-1} = \frac{V_A}{I_C}$$

The current gain is defined as  $\beta = I_C / I_B$ . The collector and the base currents are expressed as

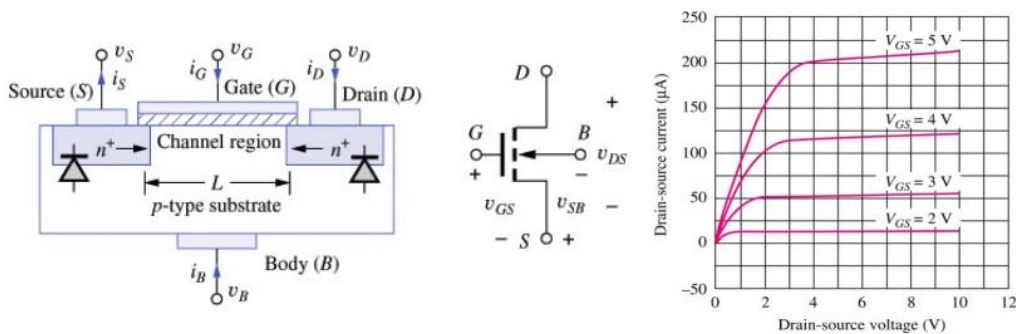
$$i_C = I_S \left[ \exp\left(\frac{v_{BE}}{V_T}\right) - \exp\left(\frac{v_{BC}}{V_T}\right) \right]$$

$$i_B = \frac{I_S}{\beta_F} \left[ \exp\left(\frac{v_{BE}}{V_T}\right) - 1 \right]$$



In semi-log plot of  $I_B$  and  $I_C$  vs.  $V_{BE}$ , both are linear curves except for very low and very high currents. The ratio between them is the current gain,  $\beta$ .

### MOSFET (Metal-oxide-semiconductor field effect transistors):



The schematic structure, symbol, and current-voltage (I-V) characteristics of an N-channel MOSFET (NMOS) are shown above. The current flowing from drain to source,  $I_{DS}$ , (or electrons moving from source to drain) is controlled by the gate-source voltage,  $V_{GS}$ . On the “flat” part of the I-V curves, the so called “saturation region”, is described by the following equation:

$$i_D = \frac{\mu_n C_{ox}}{2} \frac{W}{L} (v_{GS} - V_{TH})^2 (1 + \lambda v_{DS}) = \frac{K_n}{2} (v_{GS} - V_{TH})^2 (1 + \lambda v_{DS})$$

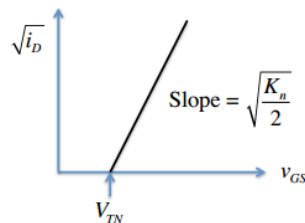
where  $V_{TH}$  is threshold voltage, and  $\lambda$  is the channel length modulation parameter.

**The parameters can be extracted experimentally:**

- Channel length modulation parameter,  $\lambda$ , can be extracted by measuring the slope of the I-V curves in the saturation region:

$$\lambda = \frac{1}{i_D} \frac{\Delta i_D}{\Delta v_{DS}}$$

- The threshold voltage and  $K_n$  can be found by plotting  $i_D^{0.5}$  vs.  $v_{GS}$  (ignore channel modulation here since  $\lambda * v_{DS} \ll 1$ ):



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